



## DESCRIPTION

The A7632A is a current mode boost DC-DC converter. Its PWM circuitry with built-in 3A current power MOSFET makes this converter highly power efficiently. The A7632A implements a constant frequency 1MHz PWM control scheme. The high frequency PWM operation also saves board space by reducing external component sizes. The A7632A features automatic shifting to pulse frequency modulation mode at light loads. Highly integration and internal compensation network minimizes as 6 external component counts. Optimized operation frequency can meet the requirement of small LC filters value and low operation current with high efficiency.

The A7632A includes under-voltage lockout, current limiting, and thermal overload protection to prevent damage in the event of an output overload.

The A7632A is available in a small SOT-26 package.

## ORDERING INFORMATION

Package Type	Part Number	
SOT-26	E6	A7632AE6R
SPQ: 3,000pcs/Reel		A7632AE6VR
Note	V: Halogen free Package R: Tape & Reel	
AiT provides all RoHS products		

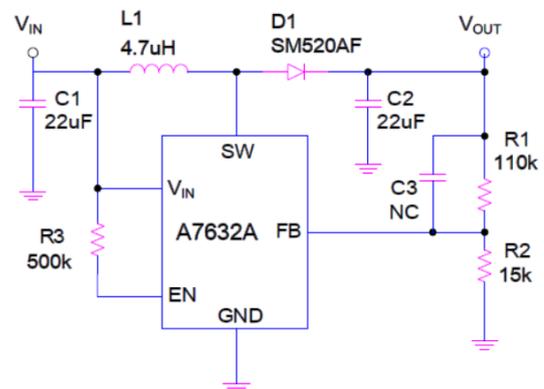
## FEATURES

- Up to 94% efficiency
- Shut-down current:1uA
- Input and Output voltage Up to 24V
- Internal Compensation, Soft-start
- 1MHz fixed frequency switching
- High switch on current:3A
- Available in SOT-26 package

## APPLICATION

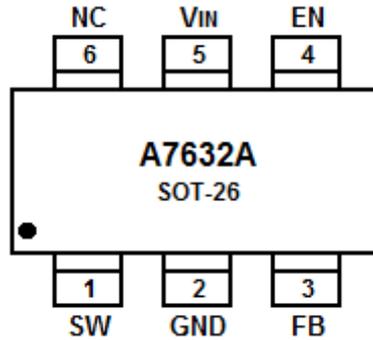
- Battery products
- Host Products
- Panel

## TYPICAL APPLICATION





## PIN DESCRIPTION



Top View

Pin #	Symbol	Function
1	SW	Switching pin.
2	GND	Ground.
3	FB	Regulation Feedback Input. Connect to an external resistive voltage divider from the output to FB to set the output voltage.
4	EN	Regulator ON/OFF Control Input. A logic high input( $V_{EN} > 1.4V$ ) turns on the regulator. A logic low input( $V_{EN} < 0.4V$ ) puts the A7632A into low current shutdown mode.
5	$V_{IN}$	Power Supply pin.
6	NC	No connecter.



## ABSOLUTE MAXIMUM RATINGS

V <sub>IN</sub> to GND	26V
SW to GND	30V
EN to GND	26V
Other Pin Voltage to GND	6V
Maximum Junction Temperature	150°C
T <sub>A</sub> , Operating Ambient Temperature Range	-40°C ~ 85°C
Maximum Soldering Temperature (at leads, 10 sec)	260°C
<b>Thermal Information</b>	
P <sub>D</sub> , Maximum Power Dissipation (T <sub>A</sub> =25°C)	0.45W
J <sub>A</sub> , Thermal Resistance	250°C/W
<b>ESD Susceptibility</b>	
HBM(Human Body Mode)	2kV

Stress beyond above listed "Absolute Maximum Ratings" may lead permanent damage to the device. These are stress ratings only and operations of the device at these or any other conditions beyond those indicated in the operational sections of the specifications are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.



## ELECTRICAL CHARACTERISTICS

$V_{IN}=3.3V$ ,  $V_{OUT}=5V$ ,  $C_{IN}=22\mu F$ ,  $C_{OUT}=22\mu F$ ,  $L=4.7\mu H$ ,  $R1=110k$ ,  $R2=15k$

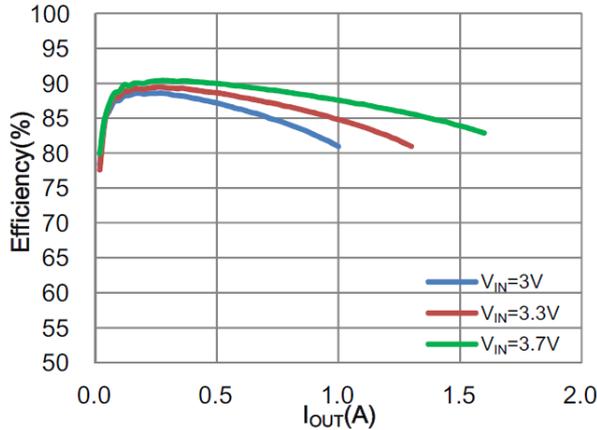
Parameter	Conditions	Min.	Typ.	Max.	Unit
Supply Voltage		2.5	-	24	V
Output Voltage Range		2.5	-	24	V
Supply Current(Shutdown)	$V_{EN}=0V$ , $V_{SW}=5V$	-	1	-	$\mu A$
Supply Current	$V_{FB}=0.7V$	-	0.35	-	mA
Feedback Voltage		0.588	0.6	0.612	V
Feedback Input Current	$V_{FB}=1.2V$	-	50	-	nA
Switching Frequency		-	1	-	MHz
Maximum Duty Cycle		85	92	98	%
EN Input Low Voltage		-	-	0.4	V
EN Input High Voltage		1.4	-	-	V
EN Input Current	$V_{EN}=5$	-	1	-	$\mu A$
Switch MOSFET Current Limit		-	3	-	A
High-Side On Resistance	$V_{OUT}=5V$	-	150	-	m $\Omega$



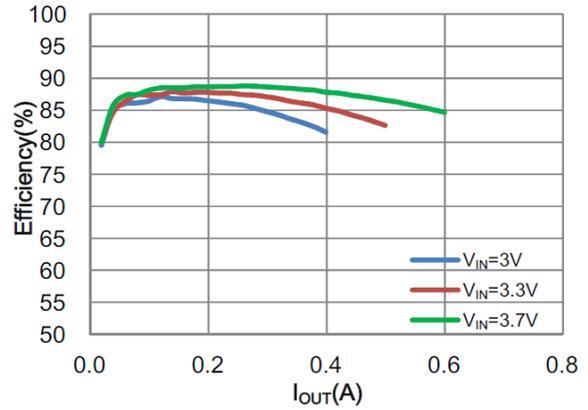
## TYPICAL PERFORMANCE CHARACTERISTICS

$C_{IN}=22\mu F$ ,  $C_{OUT}=22\mu F$ ,  $L=4.7\mu H$ ,  $T_A=25^\circ C$ , unless otherwise noted

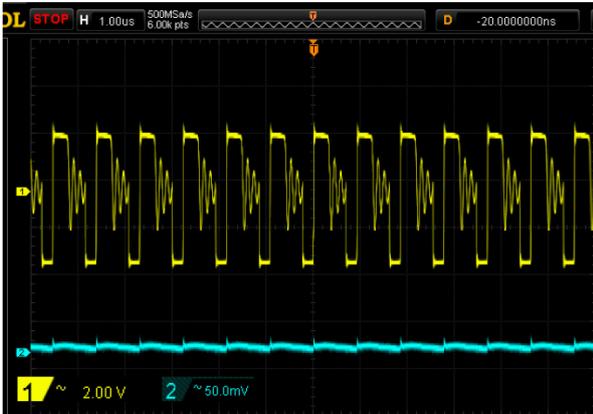
1. Efficiency vs.  $I_{OUT}$  @  $V_{OUT}=5V$



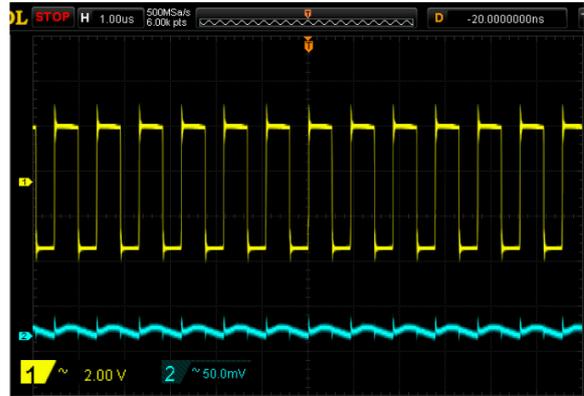
2. Efficiency vs.  $I_{OUT}$  @  $V_{OUT}=10V$



3.  $V_{IN}=3V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=50mA$ ,  
CH1= $V_{SW}$ , CH2= $\Delta V_{OUT}$



4.  $V_{IN}=3V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=200mA$ ,  
CH1= $V_{SW}$ , CH2= $\Delta V_{OUT}$



5.  $V_{IN}=3V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=500mA$ ,  
CH1= $V_{SW}$ , CH2= $\Delta V_{OUT}$

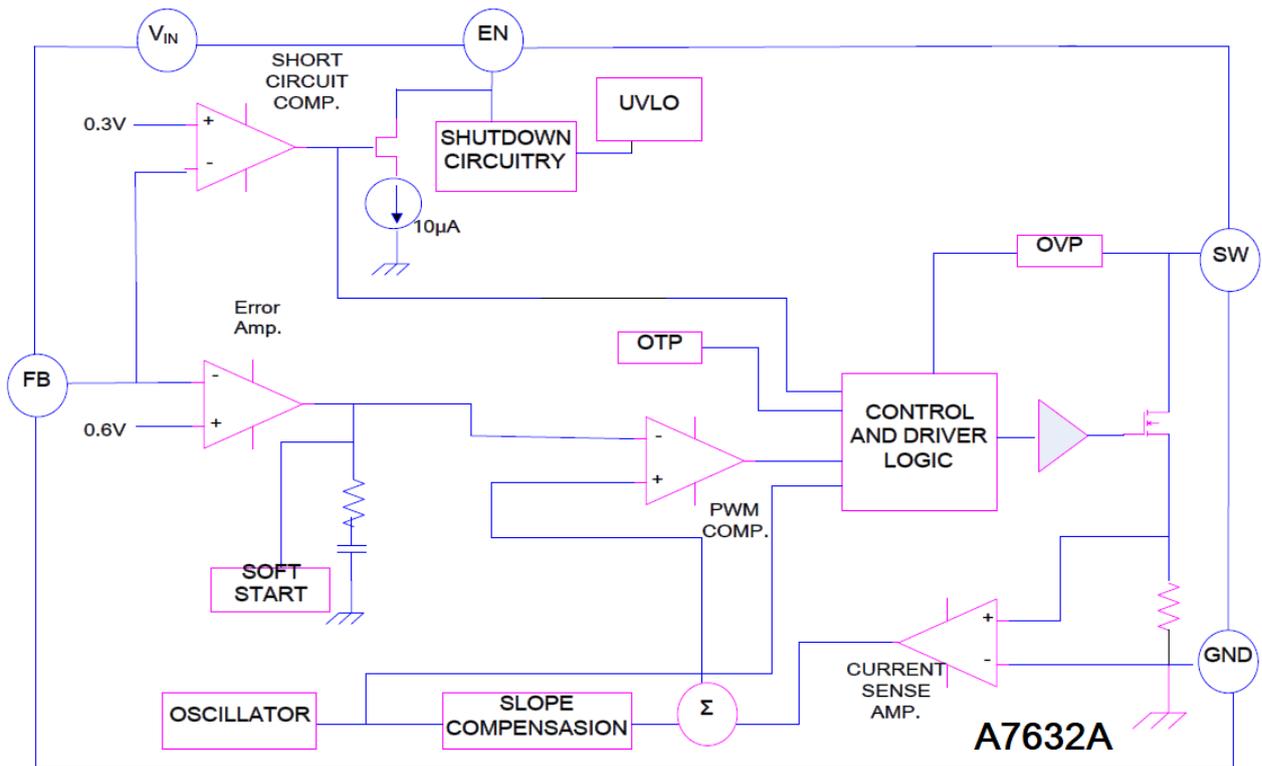


6.  $V_{IN}=3V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=1A$ ,  
CH1= $V_{SW}$ , CH2= $\Delta V_{OUT}$





**BLOCK DIAGRAM**





## DETAILED DESCRIPTION

### Operation Information

The A7632A uses a fixed frequency, peak current mode boost regulator architecture to regulate voltage at the feedback pin. At the start of each oscillator cycle the MOSFET is turned on through the control circuitry. To prevent sub-harmonic oscillations at duty cycles greater than 50 percent, a stabilizing ramp is added to the output of the current sense amplifier and the result is fed into the negative input of the PWM comparator. When this voltage equals the output voltage of the error amplifier the power MOSFET is turned off. The voltage at the output of the error amplifier is an amplified version of the difference between the 0.6V bandgap reference voltage and the feedback voltage. In this way the peak current level keeps the output in regulation. If the feedback voltage starts to drop, the output of the error amplifier increases. These results in more current to flow through the power MOSFET, thus increasing the power delivered to the output.

The A7632A has internal soft start to limit the amount of input current at startup and to also limit the amount of overshoot on the output.

### Setting the Output Voltage

Set the output voltage by selecting the resistive voltage divider ratio. The voltage divider drops the output voltage to the 0.6V feedback voltage. Use a 100k resistor for R<sub>2</sub> of the voltage divider. Determine the high-side resistor R<sub>1</sub> by the equation:

$$V_{OUT} = ( R_1 / R_2 + 1 ) \times V_{FB}$$

### Current Limitation

The internal power-MOS switch current is monitored cycle-by-cycle and is limited to the value not exceed 3A(Typ.). When the switch current reaches the limited value, the internal power-MOS is turned off immediately until the next cycle. Keep traces at this pin as short as possible. Do not put capacitance at this pin.

### Inductor Selection

For a better efficiency in high switching frequency converter, the inductor selection has to use a proper core material such as ferrite core to reduce the core loss and choose low ESR wire to reduce copper loss.

The most important point is to prevent the core saturated when handling the maximum peak current. Using a shielded inductor can minimize radiated noise in sensitive applications. The maximum peak inductor current is the maximum input current plus the half of inductor ripple current. The calculated peak current has to be smaller than the current limitation in the electrical characteristics. A typical setting of the inductor ripple current is 20% to 40% of the maximum input current. If the selection is 40%, the maximum peak inductor current is



$$I_{PEAK} = I_{IN(MAX)} + \frac{1}{2} I_{RIPPLE} = 1.2 \times I_{IN(MAX)}$$
$$= 1.2 \times \left[ \frac{I_{OUT(MAX)} \times V_{OUT}}{\eta \times V_{IN(MIN)}} \right]$$

The minimum inductance value is derived from the following equation:

$$L = \frac{\eta \times V_{IN(MIN)}^2 \times [V_{OUT} - V_{IN(MIN)}]}{0.4 \times I_{OUT(MAX)} \times V_{OUT}^2 \times f_{OSC}}$$

Depending on the application, the recommended inductor value is between 2.2μH to 10μH.

### Diode Selection

To achieve high efficiency, Schottky diode is good choice for low forward drop voltage and fast switching time. The output diode rating should be able to handle the maximum output voltage, average power dissipation and the pulsating diode peak current.

### Input Capacitor Selection

For better input bypassing, low-ESR ceramic capacitors are recommended for performance. A 22μF input capacitor is sufficient for most applications. For a lower output power requirement application, this value can be decreased.

### Output Capacitor Selection

For lower output voltage ripple, low-ESR ceramic capacitors are recommended. The tantalum capacitors can be used as well, but the ESR is bigger than ceramic capacitor. The output voltage ripple consists of two components: one is the pulsating output ripple current flows through the ESR, and the other is the capacitive ripple caused by charging and discharging.

$$V_{RIPPLE} = V_{RIPPLE\_ESR} + V_{RIPPLE\_C}$$
$$\cong I_{PEAK} \times R_{ESR} + \frac{I_{PEAK}}{C_{OUT}} \left( \frac{V_{OUT} - V_{IN}}{V_{OUT} \times f_{OSC}} \right)$$

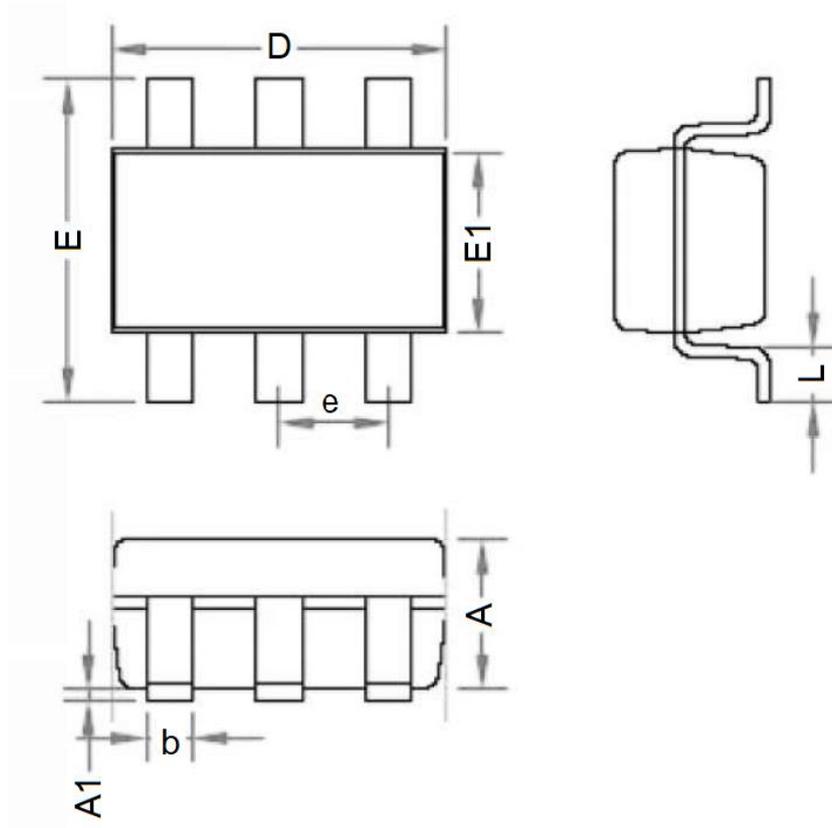
### Layout Guideline

For high frequency switching power supplies, the PCB layout is important step in system application design. In order to let IC achieve good regulation, high efficiency and stability, it is strongly recommended the power components should be placed as close as possible. The set traces should be wide and short. The feedback pin and then works of feedback and compensation should keep away from the power loops, and be shielded with a ground trace or plane to prevent noise coupling. Input and Output capacitors should be placed close to the IC and connected to ground plane to reduce noise coupling.



## PACKAGE INFORMATION

Dimension in SOT-26 (Unit: mm)



Symbols	Millimeters		Inches	
	Min.	Max.	Min.	Max.
A	-	1.45	-	0.057
A1	0.00	0.15	0.000	0.006
b	0.30	0.50	0.012	0.020
D	2.90		0.114	
E1	1.60		0.063	
e	0.95		0.037	
E	2.60	3.00	0.102	0.118
L	0.30	0.60	0.012	0.024



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